THE SQUARE-ROOT UNSCENTED KALMAN FILTER FOR STATE AND PARAMETER-ESTIMATION

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ABSTRACT

Over the last 20-30 years, the extended Kalman filter (EKF) has become the algorithm of choice in numerous nonlinear estimation and machine learning applications. These include estimating the state of a nonlinear dynamic system as well estimating parameters for nonlinear system identification (e.g., learning the weights of a neural network). The EKF applies the standard linear Kalman filter methodology to a linearization of the true nonlinear system. This approach is sub-optimal, and can easily lead to divergence. Julier et al. [1] proposed the unscented Kalman filter (UKF) as a derivative-free alternative to the extended Kalman filter in the framework of state-estimation. This was extended to parameterestimation by Wan and van der Merwe [2, 3]. The UKF consistently outperforms the EKF in terms of prediction and estimation error, at an equal computational complexity of $\mathcal{O}(L^3)^1$ for general state-space problems. When the EKF is applied to parameterestimation, the special form of the state-space equations allows for an $\mathcal{O}(L^2)$ implementation. This paper introduces the squareroot unscented Kalman filter (SR-UKF) which is also $\mathcal{O}(L^3)$ for general state-estimation and $\mathcal{O}(L^2)$ for parameter estimation (note the original formulation of the UKF for parameter-estimation was $\mathcal{O}(L^3)$). In addition, the square-root forms have the added benefit of numerical stability and guaranteed positive semi-definiteness of the state covariances.

1. INTRODUCTION

The EKF has been applied extensively to the field of nonlinear estimation for both *state-estimation* and *parameter-estimation*. The basic framework for the EKF (and the UKF) involves estimation of the state of a discrete-time nonlinear dynamic system,

$$\mathbf{x}_{k+1} = \mathbf{F}(\mathbf{x}_k, \mathbf{u}_k) + \mathbf{v}_k \tag{1}$$

$$\mathbf{y}_k = \mathbf{H}(\mathbf{x}_k) + \mathbf{n}_k, \qquad (2)$$

where \mathbf{x}_k represent the unobserved state of the system, \mathbf{u}_k is a known exogenous input, and \mathbf{y}_k is the observed measurement signal. The *process* noise \mathbf{v}_k drives the dynamic system, and the *observation* noise is given by \mathbf{n}_k . The EKF involves the recursive estimation of the mean and covariance of the state under a Gaussian assumption.

In contrast, parameter-estimation, sometimes referred to as system identification, involves determining a nonlinear mapping $\mathbf{y}_k =$

 $\mathbf{G}(\mathbf{x}_k, \mathbf{w})$, where \mathbf{x}_k is the input, \mathbf{y}_k is the output, and the nonlinear map, $\mathbf{G}(\cdot)$, is parameterized by the vector \mathbf{w} . Typically, a training set is provided with sample pairs consisting of known input and desired outputs, $\{\mathbf{x}_k, \mathbf{d}_k\}$. The error of the machine is defined as $\mathbf{e}_k = \mathbf{d}_k - \mathbf{G}(\mathbf{x}_k, \mathbf{w})$, and the goal of learning involves solving for the parameters \mathbf{w} in order to minimize the expectation of some given function of the error. While a number of optimization approaches exist (*e.g.*, gradient descent and Quasi-Newton methods), parameters can be efficiently estimated on-line by writing a new state-space representation

Wk

$$_{+1} = \mathbf{w}_k + \mathbf{r}_k \tag{3}$$

$$\mathbf{d}_k = \mathbf{G}(\mathbf{x}_k, \mathbf{w}_k) + \mathbf{e}_k, \qquad (4)$$

where the parameters \mathbf{w}_k correspond to a stationary process with identity state transition matrix, driven by process noise \mathbf{r}_k (the choice of variance determines convergence and tracking performance). The output \mathbf{d}_k corresponds to a nonlinear observation on \mathbf{w}_k . The EKF can then be applied directly as an efficient "second-order" technique for learning the parameters [4].

2. THE UNSCENTED KALMAN FILTER

The inherent flaws of the EKF are due to its linearization approach for calculating the mean and covariance of a random variable which undergoes a nonlinear transformation. As shown in shown in [1, 2, 3], the UKF addresses these flaws by utilizing a deterministic "sampling" approach to calculate mean and covariance terms. Essentially, 2L + 1, sigma points (L is the state dimension), are chosen based on a square-root decomposition of the prior covariance. These sigma points are propagated through the true nonlinearity, without approximation, and then a weighted mean and covariance is taken. A simple illustration of the approach is shown in Figure 1 for a 2-dimensional system: the left plot shows the true mean and covariance propagation using Monte-Carlo sampling; the center plots show the results using a linearization approach as would be done in the EKF; the right plots show the performance of the new "sampling" approach (note only 5 sigma points are required). This approach results in approximations that are accurate to the third order (Taylor series expansion) for Gaussian inputs for all nonlinearities. For non-Gaussian inputs, approximations are accurate to at least the second-order [1]. In contrast, the linearization approach of the EKF results only in first order accuracy.

The full UKF involves the recursive application of this "sampling" approach to the state-space equations. The standard UKF implementation is given in Algorithm 2.1 for state-estimation, and uses the following variable definitions: $\{W_i\}$ is a set of scalar weights $(W_0^{(m)} = \lambda/(L+\lambda), W_0^{(c)} = \lambda/(L+\lambda) + (1-\alpha^2+\beta),$

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 $^{^{1}}L$ is the dimension of the state variable.



Figure 1: Example of mean and covariance propagation. a) actual, b) first-order linearization (EKF), c) new "sampling" approach (UKF).

$$\begin{split} W_i^{(m)} &= W_i^{(c)} = 1/\{2(L+\lambda)\} \quad i=1,\ldots,2L). \ \lambda = \alpha^2(L+\kappa) - L \ \text{and} \ \gamma &= \sqrt{(L+\lambda)} \ \text{are scaling parameters. The constant } \alpha \\ \text{determines the spread of the sigma points around } \hat{\mathbf{x}} \ \text{and is usually} \\ \text{set to } 1e-4 \leq \alpha \leq 1. \ \kappa \ \text{is a secondary scaling parameter}^2. \ \beta \\ \text{is used to incorporate prior knowledge of the distribution of } \mathbf{x} \ \text{(for Gaussian distributions, } \beta = 2 \ \text{is optimal}). \ \text{Also note that we define} \\ \text{the linear algebra operation of adding a column vector to a matrix,} \\ i.e. \ \mathbf{A} \pm \mathbf{u} \ \text{as the addition of the vector to each column of the matrix. \ The superior performance of the UKF over the EKF has been \\ \text{demonstrated in a number of applications } [1, 2, 3]. \ \text{Furthermore,} \\ \text{unlike the EKF, no explicit derivatives } (i.e., \text{Jacobians or Hessians)} \\ \text{need to be calculated.} \end{split}$$

3. EFFICIENT SQUARE-ROOT IMPLEMENTATION

The most computationally expensive operation in the UKF corresponds to calculating the new set of sigma points at each time update. This requires taking a matrix square-root of the state covariance matrix³, $\mathbf{P} \in \mathbb{R}^{L \times L}$, given by $\mathbf{SS}^T = \mathbf{P}$. An efficient implementation using a Cholesky factorization requires in general $\mathcal{O}(L^3/6)$ computations [5]. While the square-root of \mathbf{P} is an integral part of the UKF, it is still the full covariance \mathbf{P} which is recursively updated. In the SR-UKF implementation, \mathbf{S} will be propagated directly, avoiding the need to refactorize at each time step. The algorithm will in general still be $\mathcal{O}(L^3)$, but with improved numerical properties similar to those of standard square-root Kalman filters [6]. Furthermore, for the special state-space formulation of parameter-estimation, an $\mathcal{O}(L^2)$ implementation becomes possible.

The square-root form of the UKF makes use of three linear algebra techniques⁴, *QR decomposition, Cholesky factor updating* and *efficient least squares*, which we briefly review below:

• *QR decomposition.* The QR decomposition or factorization of a matrix $\mathbf{A} \in \mathbb{R}^{L \times N}$ is given by, $\mathbf{A}^T = \mathbf{QR}$, where $\mathbf{Q} \in \mathbb{R}^{N \times N}$ is orthogonal, $\mathbf{R} \in \mathbb{R}^{N \times L}$ is upper triangu-

Initialize with:

$$\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{P}_0 = \mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T]$$
 (5)

For $k \in \{1, \ldots, \infty\}$,

Calculate sigma points:

$$\boldsymbol{\mathcal{X}}_{k-1} = \begin{bmatrix} \hat{\mathbf{x}}_{k-1} & \hat{\mathbf{x}}_{k-1} + \gamma \sqrt{\mathbf{P}_{k-1}} & \hat{\mathbf{x}}_{k-1} - \gamma \sqrt{\mathbf{P}_{k-1}} \end{bmatrix} \quad (6)$$

Time update:

1

$$\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}]$$
(7)

$$\hat{\mathbf{x}}_{k}^{-} = \sum_{i=0}^{22} W_{i}^{(m)} \mathcal{X}_{i,k|k-1}$$
(8)

$$\mathbf{P}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}]^{T} + \mathbf{R}^{\mathbf{v}}$$

$$\hat{\mathbf{y}}_{k|k-1} = \mathbf{H}[\mathbf{\mathcal{X}}_{k|k-1}]$$

$$\hat{\mathbf{y}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{Y}_{i,k|k-1}$$
(9)

Measurement update equations:

$$\mathbf{P}_{\tilde{\mathbf{y}}_{k}\tilde{\mathbf{y}}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T} + \mathbf{R}^{\mathbf{n}}$$
$$\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T} \qquad (10)$$

$$\mathcal{L}_{k} = \mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} \mathbf{P}_{\tilde{\mathbf{y}}_{k}\tilde{\mathbf{y}}_{k}}^{-1}$$
(11)

$$\hat{\mathbf{x}}_{k} = \hat{\mathbf{x}}_{k}^{-} + \mathcal{K}_{k} (\mathbf{y}_{k} - \hat{\mathbf{y}}_{k}^{-})$$
(12)

$$\mathbf{P}_{k} = \mathbf{P}_{k}^{-} - \mathcal{K}_{k} \mathbf{P}_{\tilde{\mathbf{y}}_{k} \tilde{\mathbf{y}}_{k}} \mathcal{K}_{k}^{T}$$
(13)

where $\mathbf{R}^{\mathbf{v}}$ =process noise cov., $\mathbf{R}^{\mathbf{n}}$ =measurement noise cov.



- lar and $N \geq L$. The upper triangular part of \mathbf{R} , $\tilde{\mathbf{R}}$, is the transpose of the Cholesky factor of $\mathbf{P} = \mathbf{A}\mathbf{A}^T$, *i.e.*, $\tilde{\mathbf{R}} = \mathbf{S}^T$, such that $\tilde{\mathbf{R}}^T \tilde{\mathbf{R}} = \mathbf{A}\mathbf{A}^T$. We use the shorthand notation qr{ \cdot } to donate a QR decomposition of a matrix where only $\tilde{\mathbf{R}}$ is returned. The computational complexity of a QR decomposition is $\mathcal{O}(NL^2)$. Note that performing a Cholesky factorization directly on $\mathbf{P} = \mathbf{A}\mathbf{A}^T$ is $\mathcal{O}(L^3/6)$ plus $\mathcal{O}(NL^2)$ to form $\mathbf{A}\mathbf{A}^T$.
- Cholesky factor updating. If **S** is the original Cholesky factor of $\mathbf{P} = \mathbf{A}\mathbf{A}^T$, then the Cholesky factor of the rank-1 update (or downdate) $\mathbf{P} \pm \sqrt{\nu}\mathbf{u}\mathbf{u}^T$ is denoted as $\mathbf{S} =$ cholupdate {**S**, $\mathbf{u}, \pm \nu$ }. If **u** is a matrix and not a vector, then the result is *M* consecutive updates of the Cholesky factor using the *M* columns of **u**. This algorithm (available in Matlab as cholupdate) is only $\mathcal{O}(L^2)$ per update.
- Efficient least squares. The solution to the equation

 (AA^T)x = A^Tb also corresponds to the solution of the overdetermined least squares problem Ax = b. This can be solved efficiently using a QR decomposition with pivoting (implemented in Matlab's '/' operator).

²We usually set κ to 0 for state-estimation and to 3 - L for parameter estimation [1].

³For notational clarity, the time index k has been omitted.

⁴See [5] for theoretical and implementation details.

The complete specification of the new square-root filters is given in Algorithm 3.1 for state-estimation and 3.2 for paramaterestimation. Below we describe the key parts of the square-root algorithms, and how they contrast with the stardard implementations.

Square-Root State-Estimation: As in the original UKF, the filter is initialized by calculating the matrix square-root of the state covariance once via a Cholesky factorization (Eqn. 15). However, the propagted and updated Cholesky factor is then used in subsequent iterations to directly form the sigma points. In Eqn. 19 the *time-update* of the Cholesky factor, S^- , is calculated using a QR decomposition of the compound matrix containing the weighted propagated sigma points and the matrix square-root of the additive process noise covariance. The subsequent Cholesky update (or downdate) in Eqn. 20 is necessary since the the zero'th weight, $W_0^{(c)}$, may be negative. These two steps replace the *time-update* of \mathbf{P}^- in Eqn. 8, and is also $\mathcal{O}(L^3)$.

The same two-step approach is applied to the calculation of the Cholesky factor, $\mathbf{S}_{\tilde{\mathbf{y}}}$, of the observation-error covariance in Eqns. 23 and 24. This step is $\mathcal{O}(LM^2)$, where M is the observation dimension. In contrast to the way the Kalman gain is calculated in the standard UKF (see Eqn. 11), we now use two nested inverse (or *least squares*) solutions to the following expansion of Eqn. 11, $\mathcal{K}_k(\mathbf{S}_{\tilde{\mathbf{y}}_k} \mathbf{S}_{\tilde{\mathbf{y}}_k}^T) = \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k}$. Since $\mathbf{S}_{\tilde{\mathbf{y}}}$ is square and triangular, efficient "back-substitutions" can be used to solve for \mathcal{K}_k directly without the need for a matrix inversion.

Finally, the posterior measurement update of the Cholesky factor of the state covariance is calculated in Eqn. 28 by applying Msequential Cholesky downdates to \mathbf{S}_{k}^{-} . The downdate vectors are the columns of $\mathbf{U} = \mathcal{K}_{k} \mathbf{S}_{\hat{\mathbf{y}}_{k}}$. This replaces the posterior update of \mathbf{P}_{k} in Eqn. 13, and is also $\mathcal{O}(LM^{2})$.

Square-Root Parameter-Estimation: The parameter-estimation algorithm follows a similar framework as that of the state-estimation square-root UKF. However, an $\mathcal{O}(ML^2)$ algorithm, as opposed to $\mathcal{O}(L^3)$, is possible by taking advantage of the *linear* state transition function. Specifically, the time-update of the state covariance is given simply by $\mathbf{P}_{\mathbf{w}_k}^- = \mathbf{P}_{\mathbf{w}_{k-1}} + \mathbf{R}_{k-1}^r$. Now, if we apply an exponential weighting on past data⁵, the process noise covariance is given by $\mathbf{R}_k^r = (\lambda_{RLS}^{-1} - 1)\mathbf{P}_{\mathbf{w}_k}$, and the time update of the state covariance becomes,

$$\mathbf{P}_{\mathbf{w}_{k}}^{-} = \mathbf{P}_{\mathbf{w}_{k-1}} + (\lambda_{RLS}^{-1} - 1)\mathbf{P}_{\mathbf{w}_{k-1}} = \lambda_{RLS}^{-1}\mathbf{P}_{\mathbf{w}_{k-1}}.$$
 (14)

This translates readily into the factored form, $\mathbf{S}_{w_k}^- = \lambda_{RLS}^{-1/2} \mathbf{S}_{w_{k-1}}$ (see Eqn. 31), and avoids the costly $\mathcal{O}(L^3)$ QR and Cholesky based updates necessary in the state-estimation filter. This $\mathcal{O}(ML^2)$ time update step has recently been expanded by the authors to deal with *arbitrary* diagonal noise covariance structures [7].

4. EXPERIMENTAL RESULTS

The improvement in error performance of the UKF over that of the EKF for both state and parameter-estimation is well documented [1, 2, 3]. The focus of this section will be to simply verify the equivalent error performance of the UKF and SR-UKF, and show the reduction in computational cost achieved by the SR-UKF for parameter-estimation. Figure 2 shows the superior performance of UKF and SR-UKF compared to that of the EKF on estimating the

Initialize with:

$$\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{S}_0 = \operatorname{chol}\left\{\mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T]\right\} \quad (15)$$

For
$$k \in \{1, \ldots, \infty\}$$
,

Sigma point calculation and time update:

$$\boldsymbol{\mathcal{X}}_{k-1} = \begin{bmatrix} \hat{\mathbf{x}}_{k-1} & \hat{\mathbf{x}}_{k-1} + \gamma \mathbf{S}_k & \hat{\mathbf{x}}_{k-1} - \gamma \mathbf{S}_k \end{bmatrix}$$
(16
$$\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}]$$
(17

$$\hat{\mathbf{x}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{X}_{i,k|k-1}$$
(18)

$$\mathbf{S}_{k}^{-} = \operatorname{qr}\left\{\left[\sqrt{W_{1}^{(c)}}\left(\boldsymbol{\mathcal{X}}_{1:2L,k|k-1} - \hat{\mathbf{x}}_{k}^{-}\right) \quad \sqrt{\mathbf{R}^{\mathbf{v}}}\right]\right\}$$
(19)
$$\mathbf{S}_{k}^{-} = \operatorname{cholupdate}\left\{\mathbf{S}_{k}^{-}, \quad \mathcal{X}_{0,k} - \hat{\mathbf{x}}_{k}^{-}, \quad W_{0}^{(c)}\right\}$$
(20)

$$\boldsymbol{\mathcal{Y}}_{k|k-1} = \mathbf{H}[\boldsymbol{\mathcal{X}}_{k|k-1}]$$
(21)

$$\hat{\mathbf{y}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{Y}_{i,k|k-1}$$
(22)

Measurement update equations:

$$\mathbf{S}_{\hat{\mathbf{y}}_{k}} = \operatorname{qr}\left\{\left[\sqrt{W_{1}^{(c)}}\left[\boldsymbol{\mathcal{Y}}_{1:2L,k} - \hat{\mathbf{y}}_{k}\right] \quad \sqrt{\mathbf{R}_{k}^{\mathbf{n}}}\right]\right\}$$
(23)

$$\mathbf{S}_{\tilde{\mathbf{y}}_{k}} = \text{cholupdate} \left\{ \mathbf{S}_{\tilde{\mathbf{y}}_{k}} , \ \mathcal{Y}_{0,k} - \hat{\mathbf{y}}_{k} , \ W_{0}^{(c)} \right\}$$
(24)

$$\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} = \sum_{i=0}^{22} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T}$$
(25)

$$\mathcal{K}_{k} = (\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}}/\mathbf{S}_{\tilde{\mathbf{y}}_{k}}^{T})/\mathbf{S}_{\tilde{\mathbf{y}}_{k}}$$
(26)

$$\begin{aligned} \boldsymbol{\kappa}_{k} &= \boldsymbol{\kappa}_{k} + \mathcal{K}_{k} (\boldsymbol{y}_{k} - \boldsymbol{y}_{k}) \\ \mathbf{U} &= \mathcal{K}_{k} \mathbf{S}_{\tilde{\boldsymbol{y}}_{k}} \end{aligned}$$
(27)

$$\mathbf{S}_{k} = \text{cholupdate} \left\{ \mathbf{S}_{k}^{-}, \mathbf{U}, -1 \right\}$$
(28)

where \mathbf{R}^{v} =process noise cov., \mathbf{R}^{n} =measurement noise cov.

Mackey-Glass-30 chaotic time series corrupted by additive white noise (3dB SNR). The error performance of the SR-UKF and UKF are indistinguishable and are both superior to the EKF. The computational complexity of all three filters are of the same order but the SR-UKF is about 20% faster than the UKF and about 10% faster than the EKF.

The next experiment shows the reduction in computational cost achieved by the square-root unscented Kalman filters and how that compares to the computational complexity of the EKF for parameterestimation. For this experiment, we use an EKF, UKF and SR-UKF to train a 2-12-2 MLP neural network on the well known *Mackay-Robot-Arm*⁶ benchmark problem of mapping the joint angles of a robot arm to the Cartesian coordinates of the hand. The learning curves (mean square error (MSE) vs. learning epoch) of the different filters are shown in Figure 3. Figure 4 shows how the computational complexity of the different filters scale as a function of the number of parameters (weights in neural network). While the standard UKF is $O(L^3)$, both the EKF and SR-UKF are $O(L^2)$.

⁵This is identical to the approach used in weighted recursive least squares (W-RLS). λ_{RLS} is a scalar weighting factor chosen to be slightly less than 1, *i.e.* $\lambda_{RLS} = 0.9995$.

⁶http://wol.ra.phy.cam.ac.uk/mackay

Initialize with:

$$\hat{\mathbf{w}}_0 = E[\mathbf{w}] \quad \mathbf{S}_{\mathbf{w}_0} = \operatorname{chol}\left\{E[(\mathbf{w} - \hat{\mathbf{w}}_0)(\mathbf{w} - \hat{\mathbf{w}}_0)^T]\right\} \quad (29)$$

For $k \in \{1, \ldots, \infty\}$,

Time update and sigma point calculation:

$$\hat{\mathbf{w}}_{k}^{-} = \hat{\mathbf{w}}_{k-1} \tag{30}$$

$$\mathbf{S}_{\mathbf{w}_{k}}^{-} = \lambda_{RLS}^{-1/2} \mathbf{S}_{\mathbf{w}_{k-1}} \tag{31}$$

$$\boldsymbol{\mathcal{W}}_{k|k-1} = \begin{bmatrix} \hat{\mathbf{w}}_k^- & \hat{\mathbf{w}}_k^- + \gamma \mathbf{S}_{\mathbf{w}_k}^- & \hat{\mathbf{w}}_k^- - \gamma \mathbf{S}_{\mathbf{w}_k}^- \end{bmatrix}$$
(32)

$$\mathcal{D}_{k|k-1} = \mathbf{G}[\mathbf{x}_k, \mathcal{W}_{k|k-1}]$$
(33)

$$\hat{\mathbf{d}}_{k} = \sum_{i=0}^{2D} W_{i}^{(m)} \mathcal{D}_{i,k|k-1}$$
(34)

Measurement update equations:

$$\mathbf{S}_{\mathbf{d}_{k}} = \operatorname{qr}\left\{ \left[\sqrt{W_{1}^{(c)}} \left[\boldsymbol{\mathcal{D}}_{1:2L,k} - \hat{\mathbf{d}}_{k} \right] \quad \sqrt{\mathbf{R}^{\mathbf{e}}} \right] \right\}$$
(35)

$$\mathbf{S}_{\mathbf{d}_{k}} = \text{cholupdate} \left\{ \mathbf{S}_{\mathbf{d}_{k}} , \mathcal{D}_{0,k} - \hat{\mathbf{d}}_{k} , W_{0}^{(c)} \right\}$$
(36)

$$\mathbf{P}_{\mathbf{w}_{k}\mathbf{d}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{W}_{i,k|k-1} - \hat{\mathbf{w}}_{k}^{-}] [\mathcal{D}_{i,k|k-1} - \hat{\mathbf{d}}_{k}]^{T} \quad (37)$$

$$\mathcal{K}_{k} = (\mathbf{P}_{\mathbf{w}_{k}\mathbf{d}_{k}} / \mathbf{S}_{\mathbf{d}_{k}}^{T}) / \mathbf{S}_{\mathbf{d}_{k}}$$
(38)

$$\hat{\mathbf{w}}_k = \hat{\mathbf{w}}_k^- + \mathcal{K}_k (\mathbf{d}_k - \hat{\mathbf{d}}_k)$$
(39)

$$\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\mathbf{d}_k} \tag{40}$$

$$\mathbf{S}_{\mathbf{w}_{k}} = \text{cholupdate} \left\{ \mathbf{S}_{\mathbf{w}_{k}}^{-}, \ \mathbf{U}, \ -1 \right\}$$
(41)

where \mathbf{R}^{e} =measurement noise cov (this can be set to an arbitrary value, *e.g.*, .5**I**.)

Algorithm 3.2: Square-Root UKF for parameter-estimation.



Figure 2: Estimation of the Mackey-Glass chaotic time-series (modeled by a neural network) with the EKF, UKF and SR-UKF.

5. CONCLUSIONS

The UKF consistently performs better than or equal to the well known EKF, with the added benefit of ease of implementation in



Figure 3: Learning curves for Mackay-Robot-Arm neural network parameter-estimation problem.



Figure 4: Computational complexity (flops/epoch) of EKF, UKF and SR-UKF for parameter-estimation (Mackay-Robot-Arm problem).

that no analytical derivatives (Jacobians or Hessians) need to be calculated. For state-estimation, the UKF and EKF have equal complexity and are in general $\mathcal{O}(L^3)$. In this paper, we introduced square-root forms of the UKF. The square-root UKF has better numerical properties and guarantees positive semi-definiteness of the underlying state covariance. In addition, for parameter-estimation an efficient $\mathcal{O}(L^2)$ implementation is possible for the square-root form, which is again of the same complexity as efficient EKF parameter-estimation implementations.

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