# **THE SQUARE-ROOT UNSCENTED KALMAN FILTER FOR STATE AND PARAMETER-ESTIMATION**

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## **ABSTRACT**

Over the last 20-30 years, the *extended Kalman filter* (EKF) has become the algorithm of choice in numerous nonlinear estimation and machine learning applications. These include estimating the state of a nonlinear dynamic system as well estimating parameters for nonlinear system identification (*e.g.*, learning the weights of a neural network). The EKF applies the standard linear Kalman filter methodology to a linearization of the true nonlinear system. This approach is sub-optimal, and can easily lead to divergence. Julier et al. [1] proposed the *unscented Kalman filter* (UKF) as a derivative-free alternative to the extended Kalman filter in the framework of state-estimation. This was extended to parameterestimation by Wan and van der Merwe [2, 3]. The UKF consistently outperforms the EKF in terms of prediction and estimation error, at an equal computational complexity of  $\mathcal{O}(L^3)^1$  for general state-space problems. When the EKF is applied to parameterestimation, the special form of the state-space equations allows for an  $\mathcal{O}(L^2)$  implementation. This paper introduces the *squareroot unscented Kalman filter* (SR-UKF) which is also  $\mathcal{O}(L^3)$  for general state-estimation and  $\mathcal{O}(L^2)$  for parameter estimation (note the original formulation of the UKF for parameter-estimation was  $\mathcal{O}(L^3)$ ). In addition, the square-root forms have the added benefit of numerical stability and guaranteed positive semi-definiteness of the state covariances.

#### **1. INTRODUCTION**

The EKF has been applied extensively to the field of nonlinear estimation for both *state-estimation* and *parameter-estimation*. The basic framework for the EKF (and the UKF) involves estimation of the state of a discrete-time nonlinear dynamic system,

$$
\mathbf{x}_{k+1} = \mathbf{F}(\mathbf{x}_k, \mathbf{u}_k) + \mathbf{v}_k \tag{1}
$$

$$
\mathbf{y}_k = \mathbf{H}(\mathbf{x}_k) + \mathbf{n}_k, \tag{2}
$$

where  $x_k$  represent the unobserved state of the system,  $u_k$  is a known exogenous input, and  $y_k$  is the observed measurement signal. The *process* noise  $v_k$  drives the dynamic system, and the *observation* noise is given by  $n_k$ . The EKF involves the recursive estimation of the mean and covariance of the state under a Gaussian assumption.

In contrast, parameter-estimation, sometimes referred to as system identification, involves determining a nonlinear mapping  $y_k =$ 

 $\mathbf{G}(\mathbf{x}_k, \mathbf{w})$ , where  $\mathbf{x}_k$  is the input,  $\mathbf{y}_k$  is the output, and the nonlinear map,  $\mathbf{G}(\cdot)$ , is parameterized by the vector w. Typically, a training set is provided with sample pairs consisting of known input and desired outputs,  $\{x_k, d_k\}$ . The error of the machine is defined as  $e_k = d_k - G(x_k, w)$ , and the goal of learning involves solving for the parameters <sup>w</sup> in order to minimize the expectation of some given function of the error. While a number of optimization approaches exist (*e.g.*, gradient descent and Quasi-Newton methods), parameters can be efficiently estimated on-line by writing a new state-space representation

$$
\mathbf{w}_{k+1} = \mathbf{w}_k + \mathbf{r}_k \tag{3}
$$

$$
\mathbf{d}_k = \mathbf{G}(\mathbf{x}_k, \mathbf{w}_k) + \mathbf{e}_k, \tag{4}
$$

where the parameters  $w_k$  correspond to a stationary process with identity state transition matrix, driven by process noise  $r_k$  (the choice of variance determines convergence and tracking performance). The output  $\mathbf{d}_k$  corresponds to a nonlinear observation on  $w_k$ . The EKF can then be applied directly as an efficient "secondorder" technique for learning the parameters [4].

### **2. THE UNSCENTED KALMAN FILTER**

The inherent flaws of the EKF are due to its linearization approach for calculating the mean and covariance of a random variable which undergoes a nonlinear transformation. As shown in shown in [1, 2, 3], the UKF addresses these flaws by utilizing a deterministic "sampling" approach to calculate mean and covariance terms. Essentially,  $2L + 1$ , *sigma* points (L is the state dimension), are chosen based on a square-root decomposition of the prior covariance. These sigma points are propagated through the true nonlinearity, without approximation, and then a weighted mean and covariance is taken. A simple illustration of the approach is shown in Figure 1 for a 2-dimensional system: the left plot shows the true mean and covariance propagation using Monte-Carlo sampling; the center plots show the results using a linearization approach as would be done in the EKF; the right plots show the performance of the new "sampling" approach (note only 5 sigma points are required). This approach results in approximations that are accurate to the third order (Taylor series expansion) for Gaussian inputs for all nonlinearities. For non-Gaussian inputs, approximations are accurate to at least the second-order [1]. In contrast, the linearization approach of the EKF results only in first order accuracy.

The full UKF involves the recursive application of this "sampling" approach to the state-space equations. The standard UKF implementation is given in Algorithm 2.1 for state-estimation, and uses the following variable definitions:  $\{W_i\}$  is a set of scalar weights  $(W_0^{(m)} = \lambda/(L+\lambda)$  ,  $W_0^{(c)} = \lambda/(L+\lambda) + (1-\alpha^2+\beta)$ ,

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 ${}^{1}L$  is the dimension of the state variable.



Figure 1: Example of mean and covariance propagation. a) actual, b) first-order linearization (EKF), c) new "sampling" approach (UKF).

 $W_i^{(m)} = W_i^{(c)} = 1/\{2(L + \lambda)\}\quad i = 1, \dots, 2L$ .  $\lambda = \alpha^2(L + \kappa) - L$  and  $\gamma = \sqrt{(L + \lambda)}$  are scaling parameters. The constant  $\alpha$ determines the spread of the sigma points around  $\hat{x}$  and is usually set to  $1e - 4 \leq \alpha \leq 1$ .  $\kappa$  is a secondary scaling parameter<sup>2</sup>.  $\beta$ is used to incorporate prior knowledge of the distribution of <sup>x</sup> (for Gaussian distributions,  $\beta = 2$  is optimal). Also note that we define the linear algebra operation of adding a column vector to a matrix, *i.e.*  $A \pm u$  as the addition of the vector to each column of the matrix. The superior performance of the UKF over the EKF has been demonstrated in a number of applications [1, 2, 3]. Furthermore, unlike the EKF, no explicit derivatives (*i.e.*, Jacobians or Hessians) need to be calculated.

#### **3. EFFICIENT SQUARE-ROOT IMPLEMENTATION**

The most computationally expensive operation in the UKF corresponds to calculating the new set of sigma points at each time update. This requires taking a matrix square-root of the state covariance matrix<sup>3</sup>,  $P \in \mathbb{R}^{L \times L}$ , given by  $SS^T = P$ . An efficient implementation using a Cholesky factorization requires in general  $\mathcal{O}(L^3/6)$  computations [5]. While the square-root of **P** is an integral part of the UKF, it is still the full covariance <sup>P</sup> which is recursively updated. In the SR-UKF implementation, S will be propagated directly, avoiding the need to refactorize at each time step. The algorithm will in general still be  $\mathcal{O}(L^3)$ , but with improved numerical properties similar to those of standard square-root Kalman filters [6]. Furthermore, for the special state-space formulation of parameter-estimation, an  $\mathcal{O}(L^2)$  implementation becomes possible.

The square-root form of the UKF makes use of three linear algebra techniques<sup>4</sup>, *QR decomposition*, *Cholesky factor updating* and *efficient least squares*, which we briefly review below:

 *QR decomposition.* The QR decomposition or factorization of a matrix  $\mathbf{A} \in \mathbb{R}^{L \times N}$  is given by,  $\mathbf{A}^T = \mathbf{Q} \mathbf{R}$ , where  $\mathbf{Q} \in \mathbb{R}^{N \times N}$  is orthogonal,  $\mathbf{R} \in \mathbb{R}^{N \times L}$  is upper trianguInitialize with:

$$
\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{P}_0 = \mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T] \quad (5)
$$

For  $k \in \{1, \ldots, \infty\},\$ 

Calculate sigma points:

$$
\boldsymbol{\mathcal{X}}_{k-1} = \begin{bmatrix} \hat{\mathbf{x}}_{k-1} & \hat{\mathbf{x}}_{k-1} + \gamma \sqrt{\mathbf{P}_{k-1}} & \hat{\mathbf{x}}_{k-1} - \gamma \sqrt{\mathbf{P}_{k-1}} \end{bmatrix}
$$
 (6)

Time update:

$$
\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}] \tag{7}
$$

$$
\hat{\mathbf{x}}_k^- = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{X}_{i,k|k-1}
$$
 (8)

$$
\mathbf{P}_k^- = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_k^-] [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_k^-]^T + \mathbf{R}^{\mathbf{v}}
$$

$$
\mathbf{\mathcal{Y}}_{k|k-1} = \mathbf{H}[\mathcal{X}_{k|k-1}]
$$
\n
$$
\hat{\mathbf{y}}_k^- = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{Y}_{i,k|k-1}
$$
\n(9)

Measurement update equations:

$$
\mathbf{P}_{\tilde{\mathbf{y}}_k \tilde{\mathbf{y}}_k} = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_k^-] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_k^-]^T + \mathbf{R}^{\mathbf{n}}
$$

$$
\mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_k^-] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_k^-]^T
$$
(10)

$$
\mathcal{K}_k = \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} \mathbf{P}_{\tilde{\mathbf{y}}_k \tilde{\mathbf{y}}_k}^{-1}
$$
 (11)

$$
\hat{\mathbf{x}}_k = \hat{\mathbf{x}}_k^- + \mathcal{K}_k (\mathbf{y}_k - \hat{\mathbf{y}}_k^-)
$$
 (12)

$$
\mathbf{P}_k = \mathbf{P}_k^- - \mathcal{K}_k \mathbf{P}_{\tilde{\mathbf{y}}_k \tilde{\mathbf{y}}_k} \mathcal{K}_k^T
$$
 (13)

where  $\mathbf{R}^v$ =process noise cov.,  $\mathbf{R}^n$ =measurement noise cov.



- lar and  $N \geq L$ . The upper triangular part of **R**, **R**, is the transpose of the Cholesky factor of  $P = AA^T$ , *i.e.*, **, such that**  $**R**<sup>T</sup>**R** = **AA**<sup>T</sup>$ **. We use the shorthand** notation  $qr\{\cdot\}$  to donate a QR decomposition of a matrix where only  $\tilde{R}$  is returned. The computational complexity of a QR decomposition is  $\mathcal{O}(NL^2)$ . Note that performing a Cholesky factorization directly on  $P = AA^T$  is  $\mathcal{O}(L^3/6)$ plus  $\mathcal{O}(NL^2)$  to form  $\mathbf{AA}^T$ .
- *Cholesky factor updating.* If S is the original Cholesky factor of  $P = AA<sup>T</sup>$ , then the Cholesky factor of the rank-1 update (or downdate)  $P \pm \sqrt{\nu}$ **uu**<sup>T</sup> is denoted as **S** = cholupdate  $\{S, u, \pm \nu\}$ . If u is a matrix and not a vector, then the result is  $M$  consecutive updates of the Cholesky factor using the  $M$  columns of  $\bf{u}$ . This algorithm (available in Matlab as cholupdate) is only  $\mathcal{O}(L^2)$  per update.
- *Efficient least squares.* The solution to the equation  $(AA<sup>T</sup>)x = A<sup>T</sup>$  b also corresponds to the solution of the overdetermined least squares problem  $\mathbf{A}\mathbf{x} = \mathbf{b}$ . This can be solved efficiently using a QR decomposition with pivoting (implemented in Matlab's '/' operator).

<sup>&</sup>lt;sup>2</sup>We usually set  $\kappa$  to 0 for state-estimation and to 3 – L for parameter estimation [1].

 $3$ For notational clarity, the time index k has been omitted.

<sup>4</sup>See [5] for theoretical and implementation details.

The complete specification of the new square-root filters is given in Algorithm 3.1 for state-estimation and 3.2 for paramaterestimation. Below we describe the key parts of the square-root algorithms, and how they contrast with the stardard implementations.

**Square-Root State-Estimation:** As in the original UKF, the filter is initialized by calculating the matrix square-root of the state covariance once via a Cholesky factorization (Eqn. 15). However, the propagted and updated Cholesky factor is then used in subsequent iterations to directly form the sigma points. In Eqn. 19 the *time-update* of the Cholesky factor, S , is calculated using a QR decompostion of the compound matrix containing the weighted propagated sigma points and the matrix square-root of the additive process noise covariance. The subsequent Cholesky update (or downdate) in Eqn. 20 is necessary since the the zero'th weight,  $W_0^{(c)}$ , may be negative. These two steps replace the *time-update* of  $\mathbf{P}^-$  in Eqn. 8, and is also  $\mathcal{O}(L^3)$ .

The same two-step approach is applied to the calculation of the Cholesky factor,  $S_{\tilde{y}}$ , of the observation-error covariance in Eqns. 23 and 24. This step is  $\mathcal{O}(LM^2)$ , where M is the observation dimension. In contrast to the way the Kalman gain is calculated in the standard UKF (see Eqn. 11), we now use two nested inverse (or *least squares*) solutions to the following expansion of Eqn. 11,  $\mathcal{K}_k(\mathbf{S}_{\tilde{\mathbf{y}}_k} \mathbf{S}_{\tilde{\mathbf{y}}_k}^T) = \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k}$ . Since  $\mathbf{S}_{\tilde{\mathbf{y}}}$  is square and triangular, efficient "back-substitutions" can be used to solve for  $\mathcal{K}_k$ directly without the need for a matrix inversion.

Finally, the posterior measurement update of the Cholesky factor of the state covariance is calculated in Eqn. 28 by applying <sup>M</sup> sequential Cholesky downdates to  $S_k^-$ . The downdate vectors are the columns of  $\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\tilde{\mathbf{y}}_k}$ . This replaces the posterior update of  $\mathbf{P}_k$  in Eqn. 13, and is also  $\mathcal{O}(LM^2)$ .

**Square-Root Parameter-Estimation:** The parameter-estimation algorithm follows a similar framework as that of the state-estimation square-root UKF. However, an  $\mathcal{O}(ML^2)$  algorithm, as opposed to  $\mathcal{O}(L^3)$ , is possible by taking advantage of the *linear* state transition function. Specifically, the time-update of the state covariance is given simply by  $\mathbf{P}_{\mathbf{w}_k}^{\mathbf{-}} = \mathbf{P}_{\mathbf{w}_{k-1}} + \mathbf{R}_{k-1}^{\mathbf{r}}$ . Now, if we apply an exponential weighting on past data<sup>5</sup>, the process noise covariance is given by  $\mathbf{R}_k^{\mathbf{r}} = (\lambda_{RLS}^{-1} - 1)\mathbf{P}_{\mathbf{w}_k}$ , and the time update of the state covariance becomes,

$$
\mathbf{P}_{\mathbf{w}_k}^{\top} = \mathbf{P}_{\mathbf{w}_{k-1}} + (\lambda_{RLS}^{-1} - 1)\mathbf{P}_{\mathbf{w}_{k-1}} = \lambda_{RLS}^{-1} \mathbf{P}_{\mathbf{w}_{k-1}}.
$$
 (14)

This translates readily into the factored form,  $S_{w_k}^- = \lambda_{RLS}^{-1/2} S_{w_{k-1}}$ (see Eqn. 31), and avoids the costly  $\mathcal{O}(L^3)$  QR and Cholesky based updates necessary in the state-estimation filter. This  $\mathcal{O}(ML^2)$  time update step has recently been expanded by the authors to deal with *arbitrary* diagonal noise covariance structures [7].

#### **4. EXPERIMENTAL RESULTS**

The improvement in error performance of the UKF over that of the EKF for both state and parameter-estimation is well documented [1, 2, 3]. The focus of this section will be to simply verify the equivalent error performance of the UKF and SR-UKF, and show the reduction in computational cost achieved by the SR-UKF for parameter-estimation. Figure 2 shows the superior performance of UKF and SR-UKF compared to that of the EKF on estimating the Initialize with:

$$
\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{S}_0 = \text{chol}\left\{\mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T]\right\} \quad (15)
$$

For 
$$
k \in \{1, \ldots, \infty\}
$$
,

Sigma point calculation and time update:

$$
\boldsymbol{\mathcal{X}}_{k-1} = [\hat{\mathbf{x}}_{k-1} \quad \hat{\mathbf{x}}_{k-1} + \gamma \mathbf{S}_k \quad \hat{\mathbf{x}}_{k-1} - \gamma \mathbf{S}_k]
$$
(16)  

$$
\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}]
$$
(17)

$$
\hat{\mathbf{x}}_k^- = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{X}_{i,k|k-1}
$$
\n(18)

$$
\mathbf{S}_{k}^{-} = \text{qr} \left\{ \left[ \sqrt{W_{1}^{(c)}} \left( \boldsymbol{\mathcal{X}}_{1:2L,k|k-1} - \hat{\mathbf{x}}_{k}^{-} \right) \right] \sqrt{\mathbf{R}^{\mathbf{v}}} \right] \right\} \quad (19)
$$

$$
\mathbf{S}_{k}^{-} = \text{cholupdate} \left\{ \mathbf{S}_{k}^{-} , \mathcal{X}_{0,k} - \hat{\mathbf{x}}_{k}^{-} , W_{0}^{(c)} \right\} \quad (20)
$$

$$
\boldsymbol{\mathcal{Y}}_{k|k-1} = \mathbf{H}[\boldsymbol{\mathcal{X}}_{k|k-1}] \tag{21}
$$

$$
\hat{\mathbf{y}}_k^- = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{Y}_{i,k|k-1} \tag{22}
$$

Measurement update equations:

$$
\mathbf{S}_{\hat{\mathbf{y}}_k} = \text{qr} \left\{ \left[ \sqrt{W_1^{(c)}} \left[ \mathbf{\mathcal{Y}}_{1:2L,k} - \hat{\mathbf{y}}_k \right] - \sqrt{\mathbf{R}_k^n} \right] \right\} \tag{23}
$$

$$
\mathbf{S}_{\tilde{\mathbf{y}}_k} = \text{cholupdate}\left\{\mathbf{S}_{\tilde{\mathbf{y}}_k} \ , \ \mathcal{Y}_{0,k} - \hat{\mathbf{y}}_k \ , \ W_0^{(c)}\right\} \tag{24}
$$

$$
\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} = \sum_{i=0}^{N} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T}
$$
 (25)

$$
\mathcal{K}_k = (\mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} / \mathbf{S}_{\hat{\mathbf{y}}_k}^T) / \mathbf{S}_{\hat{\mathbf{y}}_k}
$$
(26)

$$
\mathbf{x}_k = \mathbf{x}_k + \mathcal{K}_k (\mathbf{y}_k - \mathbf{y}_k)
$$
  

$$
\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\tilde{\mathbf{y}}_k}
$$
 (27)

$$
\mathbf{S}_k = \text{cholupdate} \left\{ \mathbf{S}_k^- , \mathbf{U} , -1 \right\} \tag{28}
$$

where 
$$
\mathbf{R}^{\mathbf{v}}
$$
 =process noise cov.,  $\mathbf{R}^{\mathbf{n}}$  = measurement noise cov.



Mackey-Glass-30 chaotic time series corrupted by additive white noise (3dB SNR). The error performance of the SR-UKF and UKF are indistinguishable and are both superior to the EKF. The computational complexity of all three filters are of the same order but the SR-UKF is about 20% faster than the UKF and about 10% faster than the EKF.

The next experiment shows the reduction in computational cost achieved by the square-root unscented Kalman filters and how that compares to the computational complexity of the EKF for parameterestimation. For this experiment, we use an EKF, UKF and SR-UKF to train a 2-12-2 MLP neural network on the well known *Mackay-Robot-Arm*<sup>6</sup> benchmark problem of mapping the joint angles of a robot arm to the Cartesian coordinates of the hand. The learning curves (mean square error (MSE) vs. learning epoch) of the different filters are shown in Figure 3. Figure 4 shows how the computational complexity of the different filters scale as a function of the number of parameters (weights in neural network). While the standard UKF is  $\mathcal{O}(L^3)$ , both the EKF and SR-UKF are  $\mathcal{O}(L^2)$ .

<sup>&</sup>lt;sup>5</sup>This is identical to the approach used in weighted recursive least squares (W-RLS).  $\lambda_{RLS}$  is a scalar weighting factor chosen to be slightly less than 1, *i.e.*  $\lambda_{RLS} = 0.9995$ .

<sup>6</sup>http://wol.ra.phy.cam.ac.uk/mackay

Initialize with:

$$
\hat{\mathbf{w}}_0 = E[\mathbf{w}] \quad \mathbf{S}_{\mathbf{w}_0} = \text{chol}\left\{ E[(\mathbf{w} - \hat{\mathbf{w}}_0)(\mathbf{w} - \hat{\mathbf{w}}_0)^T] \right\} \quad (29)
$$

For  $k \in \{1, \ldots, \infty\},\$ 

Time update and sigma point calculation:

$$
\hat{\mathbf{w}}_k^- = \hat{\mathbf{w}}_{k-1} \tag{30}
$$

$$
\mathbf{S}_{\mathbf{w}_k}^- = \lambda_{RLS}^{-1/2} \mathbf{S}_{\mathbf{w}_{k-1}} \tag{31}
$$

$$
\mathbf{\mathcal{W}}_{k|k-1} = \begin{bmatrix} \hat{\mathbf{w}}_k^- & \hat{\mathbf{w}}_k^- + \gamma \mathbf{S}_{\mathbf{w}_k}^- & \hat{\mathbf{w}}_k^- - \gamma \mathbf{S}_{\mathbf{w}_k}^- \end{bmatrix} \tag{32}
$$

$$
\boldsymbol{\mathcal{D}}_{k|k-1} = \mathbf{G}[\mathbf{x}_k, \boldsymbol{\mathcal{W}}_{k|k-1}] \tag{33}
$$

$$
\hat{\mathbf{d}}_k = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{D}_{i,k|k-1}
$$
 (34)

Measurement update equations:

$$
\mathbf{S}_{\mathbf{d}_k} = \mathrm{qr} \left\{ \left[ \sqrt{W_1^{(c)}} \left[ \mathbf{\mathcal{D}}_{1:2L,k} - \hat{\mathbf{d}}_k \right] - \sqrt{\mathbf{R}^{\mathbf{e}}} \right] \right\} \tag{35}
$$

$$
\mathbf{S}_{\mathbf{d}_k} = \text{cholupdate}\left\{\mathbf{S}_{\mathbf{d}_k} , \ \mathcal{D}_{0,k} - \hat{\mathbf{d}}_k , \ W_0^{(c)} \right\} \tag{36}
$$

$$
\mathbf{P}_{\mathbf{w}_k \mathbf{d}_k} = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{W}_{i,k|k-1} - \hat{\mathbf{w}}_k^-] [\mathcal{D}_{i,k|k-1} - \hat{\mathbf{d}}_k]^T
$$
 (37)

$$
\mathcal{K}_k = (\mathbf{P}_{\mathbf{w}_k \mathbf{d}_k} / \mathbf{S}_{\mathbf{d}_k}^T) / \mathbf{S}_{\mathbf{d}_k}
$$
 (38)

$$
\hat{\mathbf{w}}_k = \hat{\mathbf{w}}_k^- + \mathcal{K}_k (\mathbf{d}_k - \hat{\mathbf{d}}_k)
$$
 (39)

$$
\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\mathbf{d}_k} \tag{40}
$$

$$
\mathbf{S}_{\mathbf{w}_k} = \text{cholupdate} \left\{ \mathbf{S}_{\mathbf{w}_k}^-, \mathbf{U}, -1 \right\} \tag{41}
$$

where  $\mathbf{R}^e$ =measurement noise cov (this can be set to an arbitrary value, *e.g.*, 51.)

**Algorithm 3.2:** Square-Root UKF for parameter-estimation.



Figure 2: Estimation of the Mackey-Glass chaotic time-series (modeled by a neural network) with the EKF, UKF and SR-UKF.

#### **5. CONCLUSIONS**

The UKF consistently performs better than or equal to the well known EKF, with the added benefit of ease of implementation in



Figure 3: Learning curves for Mackay-Robot-Arm neural network parameter-estimation problem.



Figure 4: Computational complexity (flops/epoch) of EKF, UKF and SR-UKF for parameter-estimation (Mackay-Robot-Arm problem).

that no analytical derivatives (Jacobians or Hessians) need to be calculated. For state-estimation, the UKF and EKF have equal complexity and are in general  $\mathcal{O}(L^3)$ . In this paper, we introduced square-root forms of the UKF. The square-root UKF has better numerical properties and guarantees positive semi-definiteness of the underlying state covariance. In addition, for parameterestimation an efficient  $\mathcal{O}(L^2)$  implementation is possible for the square-root form, which is again of the same complexity as efficient EKF parameter-estimation implementations.

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