

## High performance InP/InAlAs/GaAsSb/InP double heterojunction bipolar transistors

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### Abstract

DC and RF characteristics of InP/InAlAs/GaAsSb/InP double heterojunction bipolar transistors (DHBTs) are reported. The device heterostructures include InAlAs spacer layer between InP emitter and GaAsSb base layer. The impact of thin InAlAs spacer layer was investigated by comparing the DC characteristics of large-area devices fabricated on InP/InAlAs/GaAsSb/InP and conventional InP/GaAsSb/InP heterostructures. By suppressing the base tunneling current and surface recombination current, the DHBTs with thin InAlAs spacer layer exhibited 5 decade lower crossover current of  $7 \times 10^{-11}$  A (collector current at unity current gain) than conventional InP/GaAsSb/InP DHBT that exhibited crossover current of  $4 \times 10^{-6}$  A. The current gain also improved twice by the impact of thin InAlAs layer. To investigate the high-frequency characteristics of InP/InAlAs/GaAsSb/InP DHBTs, small-area devices employing laterally etched undercut micro-airbridges were fabricated. The unity current gain cut-off frequency,  $f_T$ , of 100 GHz was obtained from a  $1 \times 30 \mu\text{m}^2$  emitter DHBT.

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**Keywords:** Heterojunction bipolar transistors; Spacer layer; GaAsSb; InAlAs; Compound semiconductor

### 1. Introduction

InP/GaAsSb/InP double heterojunction bipolar transistors (DHBTs) have drawn great attention for their potential applications in the area of ultra-high speed low power electronic circuits [1]. GaAsSb-based DHBTs have inherent advantages over InGaAs-based DHBTs such as simple heterostructures without complicated collector design, low base access resistance, and low surface recombination currents [2–4]. However, InP/GaAsSb/InP DHBTs suffer from high tunneling recombination current because of electron

pile up at the InP/GaAsSb emitter–base junction [5], which results in low current gain especially under low current operating conditions. The high tunneling recombination current is due to the spatially indirect type-II transition at InP/GaAsSb interface [6]. Type-II heterostructure of InP/GaAsSb junction is beneficial at a collector side, but it is detrimental at an emitter–base junction. To overcome this shortcoming of conventional InP/GaAsSb/InP DHBTs, several emitter designs have been investigated [5,7,8]. It has been reported that high quality InAlAs/GaAsSb heterostructure can be achieved much easier than InP/GaAsSb heterostructures [8]. Elegant growth techniques have been employed to achieve nearly ideal InP/GaAsSb E–B interface by utilizing MOCVD (metal organic chemical vapor

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deposition) method [9,10]. However it is very difficult to produce high quality InP/GaAsSb heterostructures by MBE (molecular beam epitaxy) so that InAlAs is a good candidate for emitter material especially for the GaAsSb-based HBTs grown by MBE. Another concern regarding aluminum containing compound semiconductor alloys grown by MBE system is high concentration of deep traps giving rise to degradation of low-frequency noise characteristics of HBTs [11,12]. It can be circumvented by avoiding the use of full InAlAs emitter.

In this study, a thin 15 nm-thick InAlAs spacer layer was inserted between the InP emitter and GaAsSb base layer to enjoy the advantages of InAlAs/GaAsSb interface and superior carrier transport properties of InP emitters with minimal effect of deep traps in InAlAs layer. The device performances of InP/InAlAs/GaAsSb/InP and conventional InP/GaAsSb/InP DHBTs were compared to investigate the impact of InAlAs spacer layer on the DC characteristics of DHBTs. We also report the RF characteristics of small-area devices ( $1 \times 30 \mu\text{m}^2$  emitter) fabricated on InP/InAlAs/GaAsSb/InP heterostructures.

## 2. Experiments

The epitaxial structures for DHBTs were grown on semi-insulating InP substrates by utilizing molecular beam epitaxy (MBE) system. Two device heterostructures used in this experiment are presented in Table 1 and band diagrams of both device heterostructures are illustrated in Fig. 1. Sample A (InP/InAlAs/GaAsSb/InP) is a DHBT heterostructure which has a 15 nm-thick InAlAs spacer layer between the InP emitter and the GaAsSb base layer. Sample B (InP/GaAsSb/InP) is a conventional GaAsSb DHBT layer which has abrupt InP/GaAsSb heterointerfaces at both E–B (emitter–base) and B–C (base–collector) junctions. To investigate the effect of the thin InAlAs

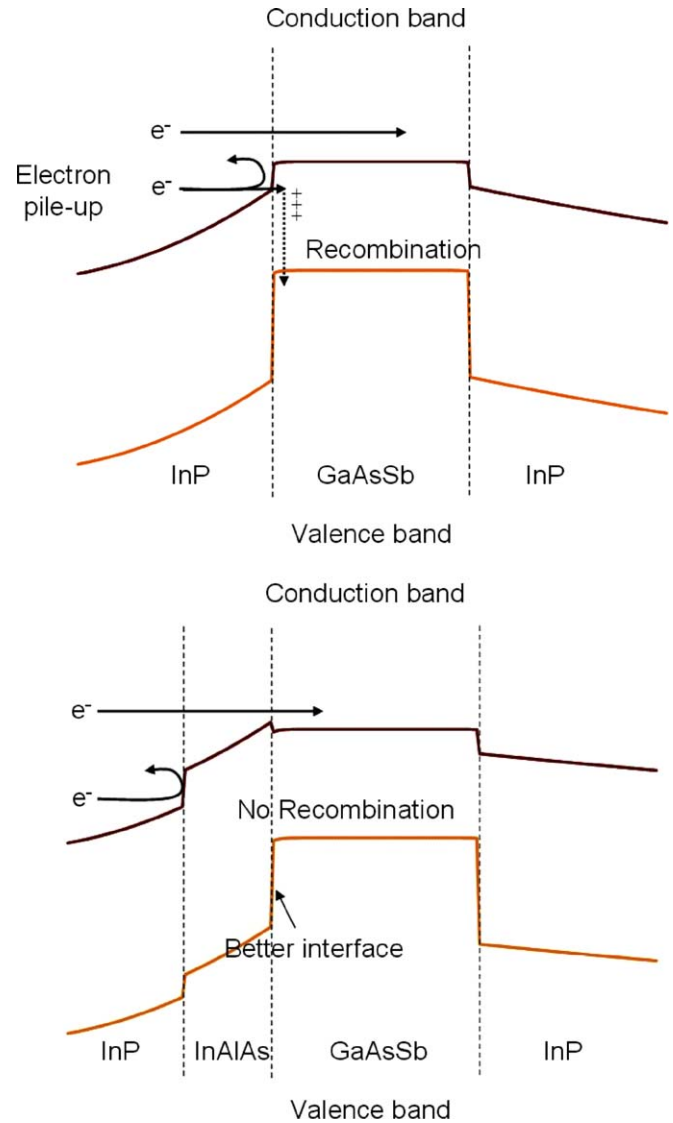


Fig. 1. The impact of InAlAs spacer layer of the InP/InAlAs/GaAsSb/InP DHBT.

Table 1  
Device heterostructures of sample A (InP/InAlAs/GaAsSb/InP) and sample B (InP/GaAsSb/InP)

	Sample A	Sample B
Emitter cap $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$	200 nm, Si: $2 \times 10^{19} \text{ cm}^{-3}$	50 nm, Si: $2 \times 10^{19} \text{ cm}^{-3}$
Emitter InP	50 nm, Si: $5 \times 10^{18} \text{ cm}^{-3}$	100 nm, Si: $4 \times 10^{18} \text{ cm}^{-3}$
Emitter InP	70 nm, Si: $4 \times 10^{17} \text{ cm}^{-3}$	70 nm, Si: $3 \times 10^{17} \text{ cm}^{-3}$
Spacer $\text{In}_{0.52}\text{Al}_{0.48}\text{As}$	15 nm, Si: $4 \times 10^{17} \text{ cm}^{-3}$	–
Base $\text{GaAs}_{0.49}\text{Sb}_{0.51}$	35 nm, C: $6 \times 10^{19} \text{ cm}^{-3}$	40 nm, C: $5 \times 10^{19} \text{ cm}^{-3}$
Collector InP	450 nm, Si: $1 \times 10^{16} \text{ cm}^{-3}$	220 nm, Si: $3 \times 10^{16} \text{ cm}^{-3}$
Collector contact $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$	30 nm, Si: $2 \times 10^{19} \text{ cm}^{-3}$	20 nm, Si: $1 \times 10^{19} \text{ cm}^{-3}$
Subcollector InP	550 nm, Si: $2 \times 10^{19} \text{ cm}^{-3}$	470 nm, Si: $4 \times 10^{18} \text{ cm}^{-3}$
S.I. InP Substrate		

spacer layer on DC characteristics of DHBTs, large-area devices with emitter sizes of  $80 \times 40$ ,  $50 \times 64$  and  $32 \times 100 \mu\text{m}^2$  were fabricated on the two DHBT structures. Wet-etch and optical lithographic techniques were utilized in the fabrication of DHBTs. The InGaAs emitter cap, GaAsSb base and InGaAs collector contact layers were etched using  $\text{H}_3\text{PO}_4\text{:H}_2\text{O}_2\text{:H}_2\text{O}$  solution. The InP emitter, InAlAs spacer, InP collector and sub-collector layers were etched with HCl-based solution. Non-alloyed ohmic metalization schemes consisting of Ti/Pt/Au and Pd/Ir/Au were used for n-type and p-type ohmic contacts. Small-area DHBTs having  $1 \times 30$  and  $2 \times 20 \mu\text{m}^2$  self-aligned emitters were also fabricated to investigate the high-frequency characteristics of InP/InAlAs/GaAsSb/InP DHBTs. The laterally etched undercut micro-airbridges were employed in the fabrication of small-area DHBTs to improve RF characteristics [13]. BCB layer was used for passivation and planarization prior to interconnecting emitter metal deposition.

### 3. Results and discussion

#### 3.1. The effects of the thin InAlAs spacer layer

The DC characteristics of the large-area devices were measured by using HP 4155B semiconductor parameter analyzer. Base sheet resistances of sample A and sample B were measured by TLM (transfer length method) were  $960 \Omega/\square$  and  $880 \Omega/\square$ , respectively. Fig. 2 shows the Gummel plots of the two types of DHBTs. There is a remarkable difference in the base currents of the two devices. The ideality factor of base current was reduced from 1.7 (sample B) to 1.26 (sample A) by introducing InAlAs spacer layer. From this observation, the dominant base current of sample A is determined to be diffusion current, while that of sample B is recombination current. The InAlAs spacer layer prevents electron pile up which occurs at type-II InP/GaAsSb E–B interface as shown in Fig. 1. By avoiding electron pile up at InAlAs/GaAsSb E–B junction, the tunneling recombination current from the emitter conduction band to the base valance band was suppressed. This effect resulted in five decade lower crossover current of  $7 \times 10^{-11}$  A (collector current at unity current gain) in comparison with that of conventional InP/GaAsSb/InP DHBT which was measured to be  $4 \times 10^{-6}$  A. Current gain also improved from 16 to 37 due to the suppressed base current in sample A. The almost same current gain characteristics were observed in the InP/InAlAs/GaAsSb/InP DHBTs with different emitter periphery to area ratios ( $80 \times 40$ ,  $50 \times 64$  and  $32 \times 100 \mu\text{m}^2$  emitter) as shown in Fig. 3. It indicates that the current gain is determined by the bulk characteristics rather than the surface characteristics of the base material [14]. The dominant surface recombination current in InP/GaAsSb/InP HBTs is direct electron injection from InP emitter sidewall to the extrinsic GaAsSb base [3]. This surface recombination current may be less severe compared in InAlAs emitter HBTs than InP emitter

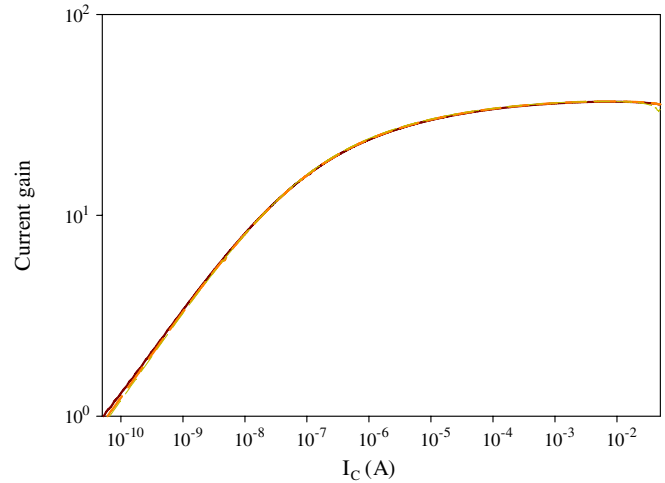


Fig. 3. The current gain as a function of collector current characteristics for sample A with various emitter periphery to area ratios (—  $A_E = 80 \times 40 \mu\text{m}^2$ , ---  $A_E = 50 \times 64 \mu\text{m}^2$ , -·-  $A_E = 32 \times 100 \mu\text{m}^2$ ).

HBTs due to the stronger surface depletion in InAlAs emitter sidewall than InP emitter sidewall.

There are two current-transport mechanisms at heterojunction interface: diffusion limitation mechanism and the conduction band barrier limitation mechanism [15]. The dominant carrier injection mechanism at E–B (emitter–base) junction of the HBTs could be identified by comparing collector currents (in forward active mode) and emitter currents (in reverse active mode) of the devices. When there is a significant conduction band discontinuity,  $\Delta E_C$ , at E–B junction, the carrier injection is dominated by tunneling or drift process and the ideality factor of the collector current in the forward active mode is much higher than unity value [16]. As shown in Fig. 4, the collector current of sample A has ideality factor of 1.07 that is very close to 1. It indicates that the current injection at emitter–base junction is dominated by diffusion process and the conduction band

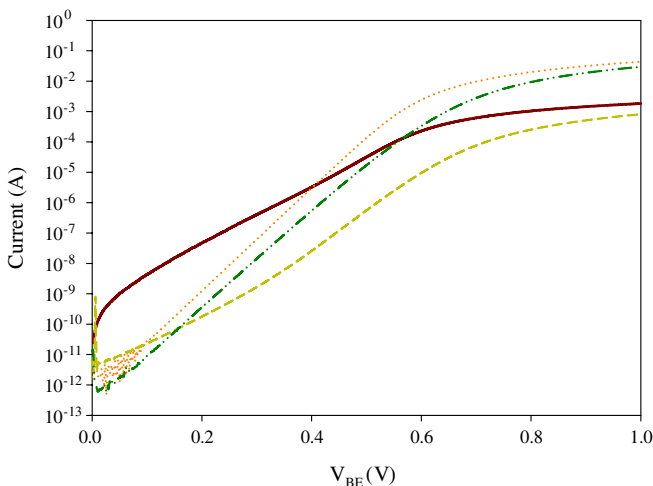


Fig. 2. Gummel plots of  $80 \times 40 \mu\text{m}^2$  GaAsSb DHBTs (—  $I_B$  of sample B, ···  $I_C$  of sample B, ---  $I_B$  of sample A, -·-  $I_C$  of sample A).

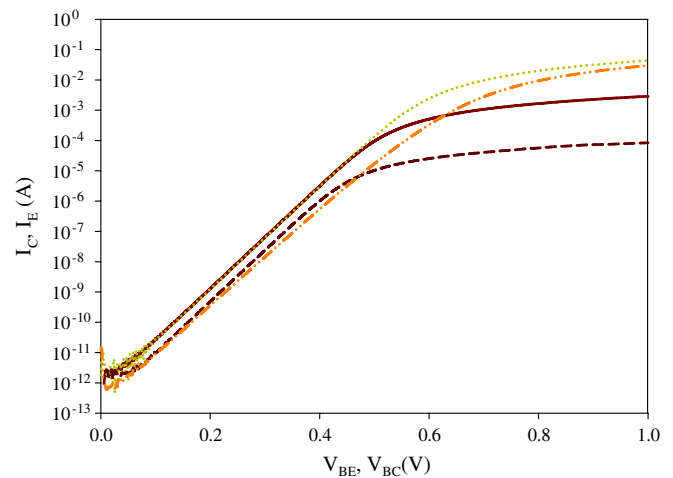


Fig. 4.  $I_C$  and  $I_E$  of  $80 \times 40 \mu\text{m}^2$  GaAsSb DHBTs in forward and reverse active modes (—  $I_E$  of sample B, ···  $I_C$  of sample B, ---  $I_E$  of sample A, -·-  $I_C$  of sample A).

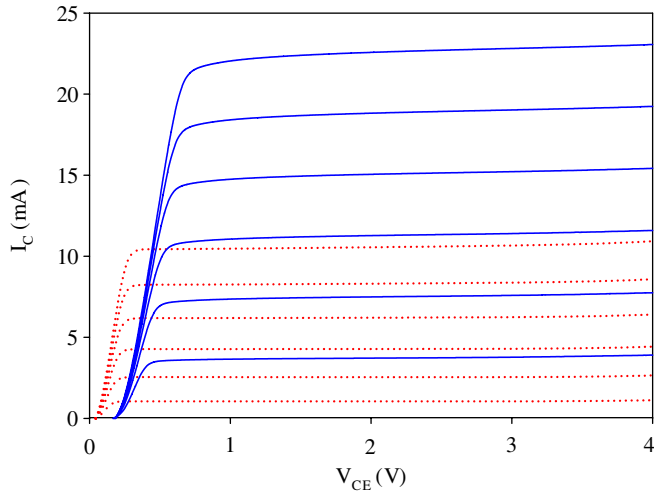


Fig. 5. Common-emitter output characteristics with base current step of 100  $\mu\text{A}$ . Emitter area is  $80 \times 40 \mu\text{m}^2$  (— Sample A; ··· Sample B).

discontinuity is insignificant for InP/InAlAs/GaAsSb E–B junction design.

As shown in Fig. 5, higher current gain was obtained in sample A which includes a thin InAlAs spacer layer. Another important parameter in DHBTs is the collector–emitter offset voltage ( $V_{\text{offset}}$ ).  $V_{\text{offset}}$  of sample A and B are 175 and 43 mV, respectively. The offset voltage can be written as:

$$V_{\text{offset}} = \left(1 - \frac{n_{\text{BC}}}{n_{\text{BE}}}\right) \cdot [V_{\text{BE}} - R_{\text{B}}I_{\text{B}}] + \frac{n_{\text{BC}}}{n_{\text{BE}}} R_{\text{E}}I_{\text{B}} + \frac{n_{\text{BC}}kT}{q} \cdot \ln \left( \frac{A_{\text{C}}J_{\text{CS}}}{A_{\text{E}}\alpha_{\text{N}}J_{\text{ES}}} \right) \quad (1)$$

where  $n_{\text{BE}}$  and  $n_{\text{BC}}$  are the ideality factors of the E–B and B–C heterojunction,  $J_{\text{ES}}$  and  $J_{\text{CS}}$  are the magnitudes of the emitter and collector saturation currents at  $V_{\text{BC}} = V_{\text{BE}} = 0$ ,  $\alpha_{\text{N}}$  is the forward current gain,  $R_{\text{E}}$  and  $R_{\text{B}}$  are emitter and base series resistances [15]. The higher  $V_{\text{offset}}$  of sample A is due to an unsymmetrical nature of the InP/InAlAs/GaAsSb/InP DHBT structure, which results in the distinct ideality factors and saturation currents in  $I_{\text{C}}$  and  $I_{\text{E}}$ . Different ideality factors and saturation currents in emitter and collector currents leads to higher  $V_{\text{offset}}$  [17,18].

### 3.2. Small-area InP/InAlAs/GaAsSb/InP DHBTs

To investigate the high-frequency characteristics of InP/InAlAs/GaAsSb/InP DHBTs, small-area DHBTs having emitter area of  $1 \times 30$  and  $2 \times 20 \mu\text{m}^2$  were fabricated. The SEM micrograph of the fabricated  $2 \times 20 \mu\text{m}^2$  DHBT with laterally etched undercut micro-airbridges is presented in Fig. 6.

Fig. 7 shows the common-emitter output characteristics of a typical  $1 \times 30 \mu\text{m}^2$  DHBT. The measured peak current gain is 32. The offset voltage and the ratio of the base–collector junction area ( $A_{\text{C}}$ ) to the emitter–base junction area ( $A_{\text{E}}$ ) have a logarithmic relationship as shown in Eq. (1). The small-area devices with large ratio of  $A_{\text{C}}/A_{\text{E}}$  resulted

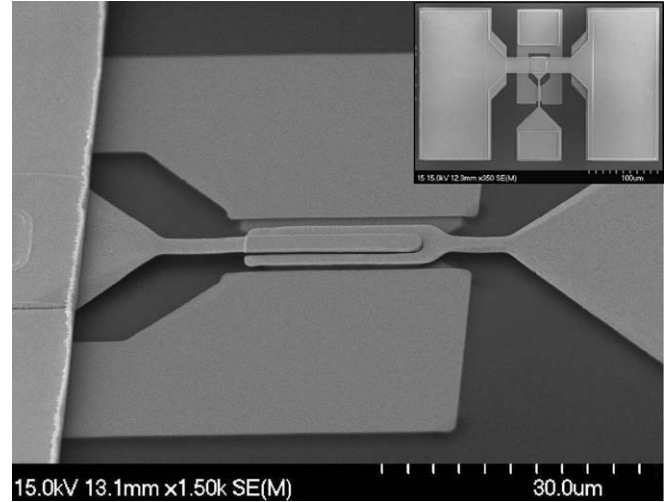


Fig. 6. SEM micrograph of the fabricated  $2 \times 20 \mu\text{m}^2$  DHBT.

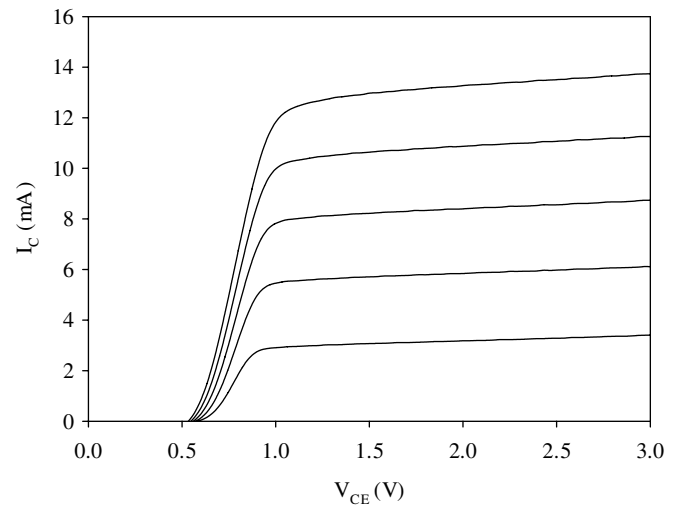


Fig. 7. Common-emitter output characteristics of  $1 \times 30 \mu\text{m}^2$  InP/InAlAs/GaAsSb/InP DHBTs with base current step of 100  $\mu\text{A}$ .

in the higher  $V_{\text{offset}}$  than the large-area devices although both devices had the same InP/InAlAs/GaAsSb/InP heterostructure. A high breakdown voltage of 6.6 V ( $BV_{\text{CEO}}$ ) was measured for the  $1 \times 30 \mu\text{m}^2$  emitter InP/InAlAs/GaAsSb/InP DHBTs whose breakdown voltage was defined at collector current of 100  $\mu\text{A}$ .

On-wafer microwave  $S$ -parameter measurements were performed on InP/InAlAs/GaAsSb/InP DHBTs at frequencies up to 40 GHz, using an HP8510C vector network analyzer. Fig. 8 shows the common emitter short circuit current gain ( $h_{21}$ ) of the typical small-area ( $1 \times 30$  and  $2 \times 20 \mu\text{m}^2$ ) DHBTs. Current gain curves were extrapolated with a  $-20$  dB/decade roll-off to obtain unity current gain cut-off frequency,  $f_{\text{T}}$ . Unity current gain cut-off frequencies were measured to be 60 and 100 GHz for  $2 \times 20$  and  $1 \times 30 \mu\text{m}^2$  emitter DHBTs, respectively. For the maximum  $f_{\text{T}}$  values, the transistors were biased at  $V_{\text{CE}} = 2.5$  V and 2 V and  $I_{\text{C}} = 12.2$  mA and 18.5 mA for  $2 \times 20$  and

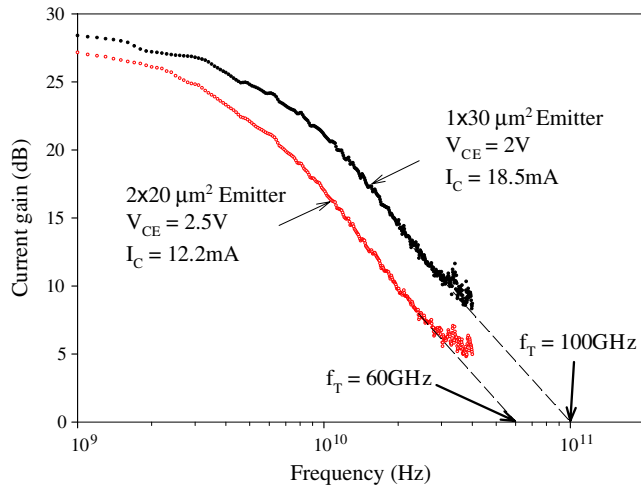


Fig. 8. Frequency dependence of  $h_{21}$  for InP/InAlAs/GaAsSb/InP DHBTs with  $1 \times 30$  and  $2 \times 20 \mu\text{m}^2$  emitter sizes.

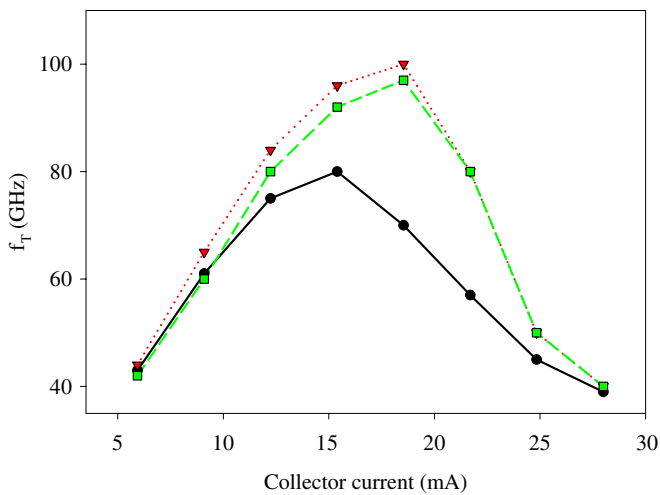


Fig. 9. Common emitter current-gain cut-off frequency ( $f_T$ ) versus collector current at various  $V_{CE}$  for  $1 \times 30 \mu\text{m}^2$  InP/InAlAs/GaAsSb/InP DHBTs. (—●—  $V_{CE} = 1$  V, —▼—  $V_{CE} = 2$  V, —■—  $V_{CE} = 3$  V).

$1 \times 30 \mu\text{m}^2$  emitter DHBTs, respectively. A peak maximum oscillation frequency,  $f_{MAX}$ , was also extracted from the measured  $S$ -parameter. A relatively low  $f_{MAX}$  of 62 GHz for  $1 \times 30 \mu\text{m}^2$  emitter DHBTs may be ascribed to the large base–collector parasitic capacitance  $C_{BC}$ , which can be improved by undercut etching of collector layer and further scaling of base width.

The variation of  $f_T$  with collector current for the  $1 \times 30 \mu\text{m}^2$  emitter InP/InAlAs/GaAsSb/InP DHBTs was measured at  $V_{CE} = 1, 2$  and  $3$  V as shown in Fig. 9. A peak  $f_T$  of 100 GHz is obtained at collector of 18.5 mA. When the collector current was further increased beyond 18.5 mA,  $f_T$ 's was degraded. The reduced  $f_T$  at high collector current can be ascribed to Kirk effect [19].

#### 4. Conclusion

The performances of InP/InAlAs/GaAsSb/InP and InP/GaAsSb/InP DHBTs were compared and the impact of the

thin InAlAs spacer layer was investigated. The InP/InAlAs/GaAsSb/InP DHBTs exhibited the improved DC characteristics by effectively suppressing tunneling recombination at E–B junction. The DC current gain of 37 and very low crossover current of  $7 \times 10^{-11}$  A were demonstrated. The conduction band discontinuity,  $\Delta E_C$ , at InAlAs/GaAsSb interface was also found out to be negligible. The microwave performance of InP/InAlAs/GaAsSb/InP DHBTs was also characterized. The current-gain cut-off frequency of 100 GHz was obtained for  $1 \times 30 \mu\text{m}^2$  emitter DHBTs.

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