

A STATE OF THE ART ON ADC MODELING

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Abstract - *The state of the art of the research on modelling of analog-to-digital converter-based measuring devices is surveyed. Main topics of modelling are reviewed according to the fields of prevailing scientific interest in metrological research such as quantization models, error models, and correction-aimed models. In these fields, recent developments are analysed with the aim of focusing both the contemporary situation and the imminent trends.*

1. INTRODUCTION

In the wider and wider development of analog-to-digital conversion systems, a pre-eminent role has been played by modelling techniques.

Modelling of analog-to-digital converter (ADC) components, as well as of digital measuring systems based on ADCs, allows the device behaviour to be predicted with a few of preventive experiments. For this reason, in last years a great deal of scientific interest has been directed to ADC modelling also for metrological aims. A model turns out to be useful for investigating the ADC metrological behaviour in several operating conditions during the main phases of development: *design*, *evaluation*, and *improvement*. In ADC *design*, the pre-eminent intrinsic error source is the quantization, and theoretical fundamental studies have been devoted to this topic by specialised research groups. In ADC *evaluation*, modelling is mainly used to analyse the impact of error sources on the metrological behaviour. In ADC *improvement*, the error occurrence is predicted in a range of operating conditions as wide as possible in order to compensate error source effects and correct deterministic errors. In all these fields, a lot of research activities with fruitful developments are in process.

In this paper, recent developments and current trends which are focus of prevailing scientific interest on ADC modelling are analysed by referring to: (i) *quantization models*, (ii) *error models*, and (iii) *correction-aimed models*.

2. QUANTIZATION MODELS

In last ten years, research on ADC modelling related to the quantization can be classified on the basis of the effects of quantization models on (i) the *ADC errors*, (ii) the *dithering*, and (iii) the *testing strategy*.

In the last decade, the effects of quantization models on *ADC errors* have been investigated with the main aim of assessing the conditions for the model validity [1]-[3]. Conversely, the current research trend is aimed at overcoming the limitations due to the assumption of uniform distribution of the quantization noise, and at investigating the quantization error for non ideal quantizers [4]-[6]. In particular, for a uniform [8], nonuniform [9], nonmonotonic and hysteretic quantizer [10], the probability density function (*pdf*) and the variance of the quantization error was derived.

The effects of quantization models on *dithering* have been investigated by referring to the well known idea of dithering which consists of "whitening" the autospectral density of the quantization noise by adding a suitable signal, and then reducing this noise by digital signal processing [11]-[14]. This means that a significant portion of the error power is located outside the signal band and can be eliminated by low-pass filtering. A sound theoretical background for the dithering theory was presented by Widrow et al. in their survey paper on quantization [3]. ADC linearity is usually increased through wide-amplitude dithering. A useful tool was provided by deriving relationships for the quantization error in the case of discrete binary, uniform, and Gaussian dither signals [15]. Then, a figure of merit ("D") was introduced as the deviation of the ADC characteristic function from the unity gain line [16]- [17]. In last developments, for both additively and subtractively-dithered ADCs theoretical formulae were derived. [18]. Finally, the problem of designing dither-based quantizing system was addressed by giving quantitative criteria for choosing the parameters acting on resolution and accuracy [19], and for quasi-static signals corrupted with network induced interference or normally distributed white noise [12]-[14].

A *testing strategy* for ADCs necessarily implies a test signal model and a quantizer error model [20]. In

particular, different model choices influence significantly the test results. A first case relates to the offset of the test signal. In [21], the well known results for null offset were generalised to the case of unknown offset by deriving the expected value and variance of the noise versus the offset and the number of levels. Besides to the offset, the amplitude of the test signal was also investigated [22]. Quantization error modelling is also very important in testing strategy. As an example, the zero-memory quantizer model loses completely significance in the case of high-speed flash ADCs [23]. The inadequacy of the reference quantizer theory behind the IEEE 1057 standard sine wave test in the case of actual nonlinear quantizers was pointed out in [6]. This situation was shown to be significantly improved by the above mentioned generalisation of the Widrow's quantization theorem to a generic quantizer (nonuniform and/or nonmonotonic).

3. ERROR MODELS

Research on ADC modelling includes predominantly a lot of works carried out in the framework of ADC design not strictly devoted to metrological aims and thus out of the scope of this paper. However, also in design, considerations on metrological performance are not fully neglected. Moreover, apart from their final aims, in some cases ADC models are strongly influenced by the conversion mechanism and, particularly, by the influence of peculiar error sources inside the architecture. Consequently, in the following, at first some general criteria and categories for the *classification of the ADC error models* are at first given. Then, on this basis, the ADC error modelling techniques recent focus of scientific interest are reviewed according to their dependence on the conversion mechanism as *architecture-independent*, and *architecture-dependent*.

3.1 Classification

ADC error modelling is generally approached at *analytical* or *heuristic* level. *Analytical* models include knowledge on the ADC error mechanism and conversion principle defined through mathematical relations and/or procedures. They are classified according to the abstraction level [24]-[25]: (i) *electrical models*, (ii) *macromodels*, and (iii) *behavioural models*. *Electrical models* details the ADC metrological behaviour at level of electronic components typical of the architecture. This allows the error effects to be finely tuned by the designer by analysing the influence of each electronic component of

the architecture circuitry. Conversely, they are not easily utilisable in all those applications where all the details on the electrical behaviour are not needed owing to the high computing times. *Macromodels* analyse the ADC behaviour through electrical equivalent circuits simpler than the actual ones (such as e.g. Thevenin or Norton-based). Such an approach is limited by the practical difficulty of finding a suitable equivalent topology without losing significant information on the real ADC metrological behaviour. *Behavioural models* do not take account the physical realisation of the ADC at all. The device is characterised by input-output analytical or numeral relations, without going in deep into the internal structure. Behavioural models are further classified according to their flexibility in [26-28]: (i) *table models*, (ii) *explicit models*, and (iii) *implicit models*. *Table models* memorises the input-output characteristic in a look-up table. This strategy turns out to be effective in the verification generally following the design, but is rigid because depends on the specific ADC under analysis: each aleatory variation of model parameters requires a new table generation, i.e. minimum flexibility. Moreover, the look-up table implementation requires significant times and memory space. *Explicit models* describe the ADC behaviour by an analytical relation in a closed form easy to be represented in a programming language. However, this consists also in their main limit because easy software packages are not always immediately available. Main advantage is the possibility of introducing in the analytical relation suitable parameters to describe different working conditions. This makes flexible the model structure, though the specific ADC architecture is not easy to be left out of consideration. *Implicit models* characterises the converter either in the time or frequency domain, by differential equations with suitable parameters to account for different architectures as well as device classes. They allows the maximum flexibility in the description of the ADC behaviour because they can be oriented to the design of a specific converter, a particular architecture. More in general, a whole class of devices can be described by parametrising the model as a whole and by assigning from time to time the parameter values according to the ADC under analysis. By creating a library of such a model, this allows a simulation environment to be created for verify and optimise the metrological performance of a Whatever ADC in all the design cycle. Finally, in several cases, electrical-behavioural mixed models are utilised. They describe some ADC sections in terms of electrical models while other as behavioural models. This allows

the description detail of the error effects to be tailored directly in the more interesting specific sections.

Heuristic models include and describe knowledge on ADC error sources and conversion mechanism having empirical character (e.g. human skill), and/or particular nonlinear patterns (e.g. two-dimensional error frames). In particular, Artificial Neural Networks (ANNs) have found fruitful applications in ADC error modelling owing to their capability of successfully modelling complex nonlinear behaviours [29]. Behavioural modelling limitations due to either or the particular ADC architecture [30]-[31], or to the used parameter identification technique [32] were overcome by ANNs with ease of use, generalisation capabilities, and usefulness of obtainable results. The neural modelling approach is based on the use of an ANN which is capable, after a proper set-up phase, of providing an output digital code corresponding to the one that can be obtained from the actual ADC to be modelled. The approach proposed in [29], [33] is architecture-independent and uses the well known identification techniques based on ANNs [34].

Apart from the knowledge definition, the modelling approach can be "*a-priori*" or "*a-posteriori*" [25], [35]: the former exploits available information such as on the architecture and/or on the conversion principle, whereas the latter utilises only experimental output data.

The choice of the modelling approach depends on the use of the model: as an example, an electrical model can be fruitfully used in ADC design, whereas a behavioural model allows the ADC to be simulated as a component in the design of a more complex system such as a digital measurement system. In any case, a general criterion for selecting the modelling approach is based on the trade-off between the model accuracy and the corresponding simulation burden.

3.2 Architecture-Independent Error Models

A classification on architecture-independent error models can be carried out also according to the *static* or *dynamic* nature of the ADC input, and consequently to the ADC error nature. Static models characterise the

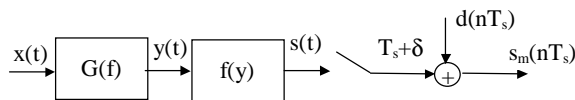


Fig.1 ADC behavioural error model including the major error mechanisms ($x(t)$: input signal, $G(f)$ dynamic errors of the transfer function, $f(y)$: static nonlinear distortions, d : time jitter, $s(t)$: actual signal before sampling and quantization, T_s : sampling period, and $d(nT_s)$: additive noise (random+spurious) and quantization errors, $s_m(nT_s)$: actual digital output signal).

converter for constant or slowly variable input signals, whereas dynamic models are used for higher frequency input signals.

The most general and simple *static error model* was developed at behavioural level with the aim of describing the ADC nonlinearity [36]. This static memory-less approach [35] was followed also by other authors with the aim of describing the nonlinear transfer characteristic of the ADC. A more accurate model turns out to be also more architecture-oriented. An example though still general in its concept is the behavioural model proposed by Ruan for the three more diffused ADC architectures (successive approximation, dual slope and flash) [37]. Finally, the problem of the harmonic noise due to the nonlinearity of the transfer function of an actual ADC was faced by the analytical modelling of INL (integral nonlinearity) through simple power functions [38]. An investigation on the influence of architecture on the static modeling strategy is reported in [39].

Research on dynamic modeling is really reach and promising. It can be classified as:

- (i) *jitter models*: research in the last decade was mainly aimed at investigating error effects and its bounds [40]-[42]. These results were derived in the most general case of lack of synchronisation between signal generator and data acquisition. However, they can not applied in all those practical cases of synchronised measurement systems (signal generation + data acquisition). In this case, Schoukens studied the influence of time jitter on the error by deriving explicit expressions for the related measurement errors [43]. Stenbakken in its jitter testing discussion used two different models: one based on a sawtooth wave and another based on a sine wave superimposed to a ramp, by discussing they adequacy for a digital scope [44].
- (ii) *models of the actual acquisition channel*: in this field, in the last decade, research was mainly aimed at deriving a model of the actual data acquisition channel as a whole; in the years, the model was progressively complicated and made more robust from the theoretical and identification point of view. Further error sources, such as time base distortion, amplitude nonlinearity, were added, by mainly following a theoretical approach of parameter estimation and system identification (Fig.1) [43],[45]-[52]. A proposal of standardizing the modelling, identification, and optimisation was advanced [53].
- (iii) *nonlinearity models*: a significant research effort was devoted to the model the ADC nonlinearity. Several and different approaches were

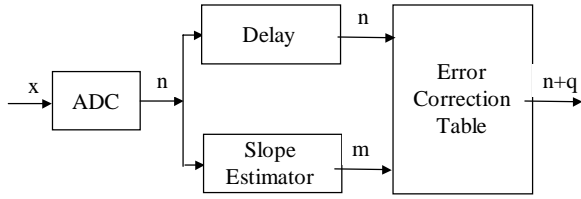


Fig.2 Block diagram of the phase-plane compensation architecture (x : ADC input, n : ADC uncorrected output code bin, m : estimated slope of the input signal, q : error correction).

devoted to the challenge of deriving mathematical relationships between distortion, input signal parameters and ADC nonlinearity errors [44]-[60].

- (iv) *statistical models*: another approach was aimed at describing the device behaviour via a suitable statistical methodology [61]-[64]. As an example, the transfer characteristic of the ADC is described via a suitable conditional probability function, estimated through a modified version of the popular histogram test.
- (v) *software models*: especially at electronic manufacturer level, a significant contribution on modelling research was devoted for describing as much accurately as possible the actual dynamic behavior of ADC devices. This was necessary to the designer for diagnostic and evaluation purposes in the design phase. Specific software simulation tools were developed both at behavioural and at electrical-circuit level [65]-[68].

3.3 Architecture-dependent error models

Low-level modelling proposals reported in literature are necessarily architecture-dependent. Owing to their extreme working conditions, those devoted to such as *sigma-delta ADCs* will be remarked.

As above remarked in the general case of ADCs, also for *sigma-delta ADCs* a great effort has been spent by designer in order to improve dynamic performance through modelling [69]-[70]. However, modelling specifically aimed at analysing error effects with metrological scope only recently has become focus of scientific interest owing to the advent of new high-accuracy sigma-delta ADCs for instrumentation [71]. Model proposals specifically aimed at metrological aims. are referred to the analysis of the main error sources directly inside the sigma-delta ADC architecture, and (ii) for their effects on the overall metrological performance [72]-[77].

4. CORRECTION-AIMED MODELS

Error modelling has played a main role also in generating a correction of the ADC actual metrological

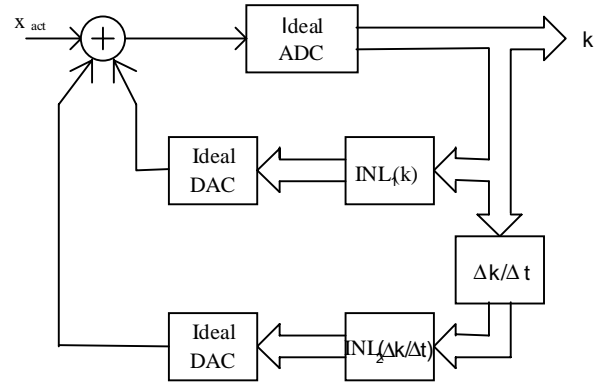


Fig.3 Block diagram of the a priori phase-plane error compensation architecture (x_{act} : actual input, INL_1 and INL_2 : static and dynamic additive components).

behaviour. Various solutions were proposed either (i) independently of the architecture in order to maximise generality, or (ii) exploiting the architecture peculiarities, in order to maximise effectiveness.

Architecture-independent strategies were based initially on an on-line correction obtained by subtracting the modelled dynamic error to each output sample of the actual ADC [35]-[36],[78]-[79]. The addition of the time slope of the input signal as a second dimension allowed varies frequencies and amplitude for a larger class of input signals to be discriminated (Fig. 2) [36]. In this way, the ADC error is described as a function of the output code as well as the time slope of the input, i.e. in the "phase-plane". This permits the modelling of both the in-phase and in-quadrature phase distortions closer to actual ADCs, having memory behaviour or not real transfer function. Investigation on the possibility of adding a third dimension for modelling second-order derivative effects demonstrated that error is mainly dominated by first-order effects [80]. The ADC correction based on phase-plane modelling showed to work adequately in several cases [81]-[85]. Apart from improvements strictly related to correction mechanism [86], phase plane modelling was improved mainly in model identification, and, namely, in: (i) the calibrating signals [87]-[89], and (ii) the experimental burden [83], [85], [87], [90]. However, such a modelling approach, completely a-posteriori in its concept, has main limitation just in its generality: the error model needs for its identification a burdensome experimental work. An alternative architecture-dependent idea has been based on an analytical a-priori approach to phase-plane modelling for most popular ADC architectures (integrating, successive-approximation, and flash ADCs), both for the static [39] and for the dynamic case (Fig. 3). Knowledge on the error source action inside the ADC architecture is exploited to

mathematically derive a peculiar shape of the phase-plane surface [91]-[92].

5. CONCLUSIONS

The state of the art and the leading trends of the research in the field of ADC modeling have been discussed. The paper is aimed at providing young researchers interested in ADCs with a helping tool for orientating and more quickly become effective in this research field.

The capability of this field of attracting the scientific focus of interest is testified by the large amount of scientific contributions devoted by high-level research centres active in the field from several years.

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